



EuCAP 2007
Nov. 12-16, 2007

Multi-Keyhole Model for MIMO Radio-Relay Systems

Y. KARASAWA, M. Tsuruta and T. Taniguchi

Advanced Wireless Communication research Center (AWCC)
Univ. Electro-Communications (UEC Tokyo)

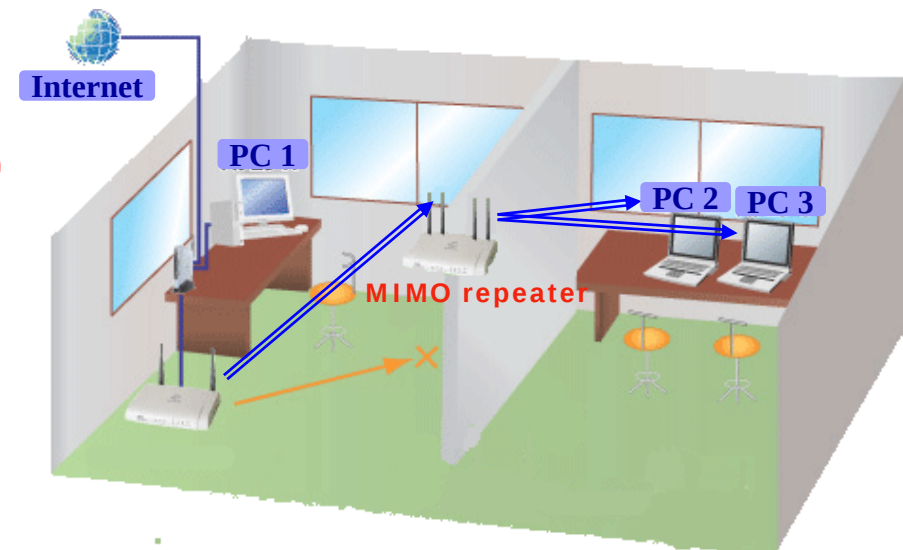


Outline of Presentation

1. Radio-relay system for MIMO service area expansion
2. Multi-keyhole model for the system performance evaluation
3. Empirical formula for PDF of the largest eigenvalue
4. Empirical formula for PDFs of all eigenvalues
5. Evaluation of the estimation accuracy
6. Conclusions

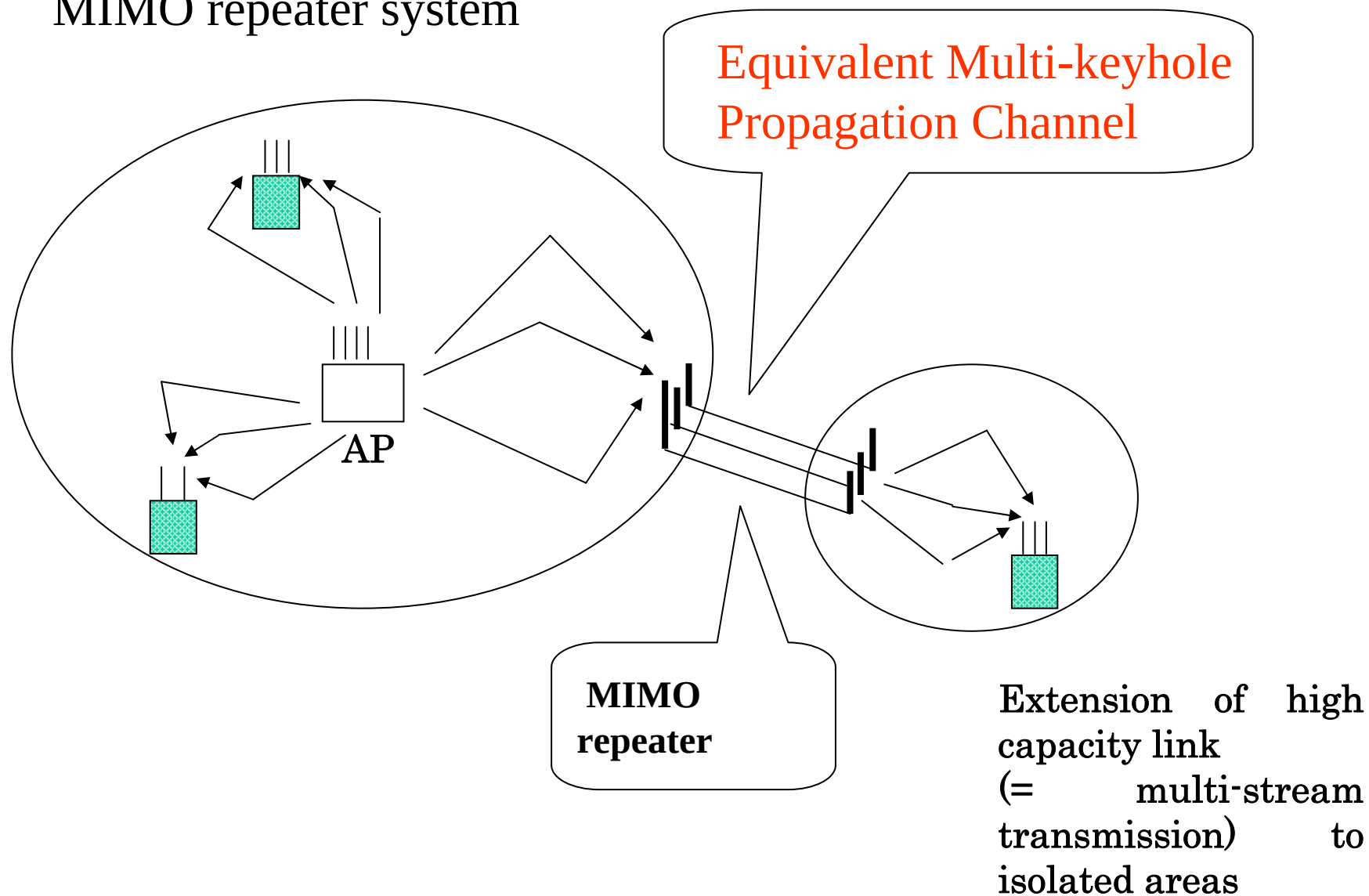
Motivation

- Why do we need MIMO channel models?
 - High data rate transmission (WLAN, WiMAX)
- Service expansion to isolated areas
 - MIMO repeater system
(MIMO radio-relay system)
 - Ad Hoc Network



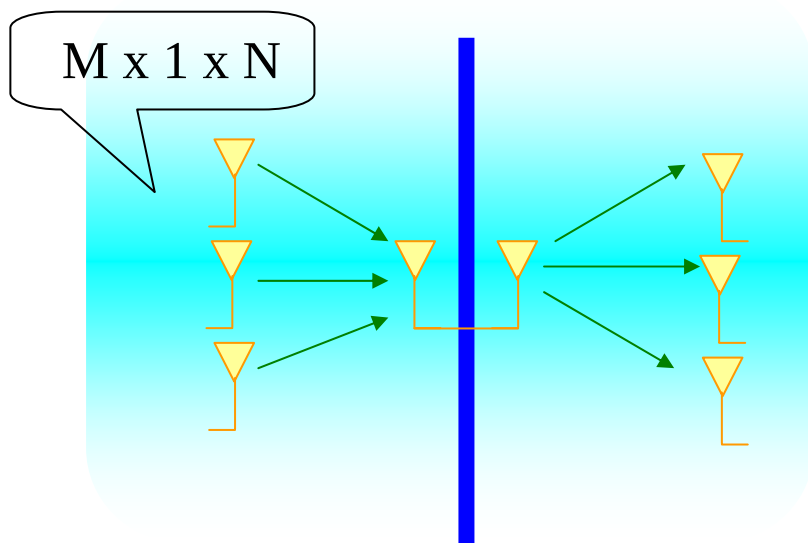


MIMO repeater system

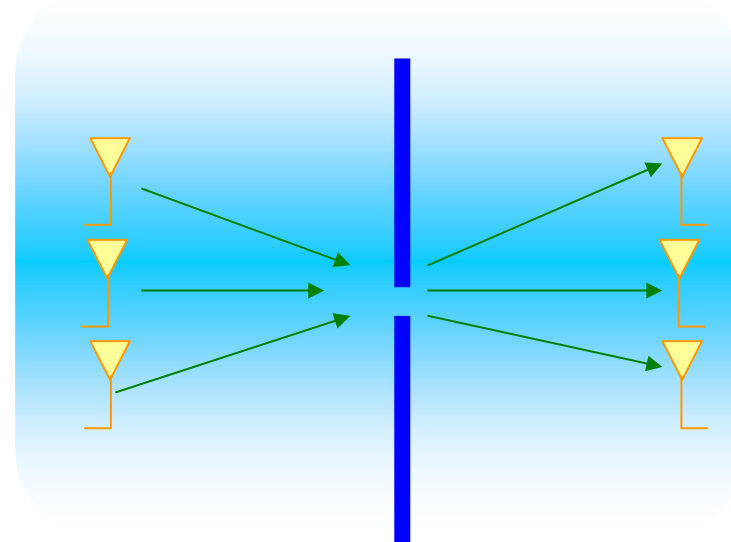




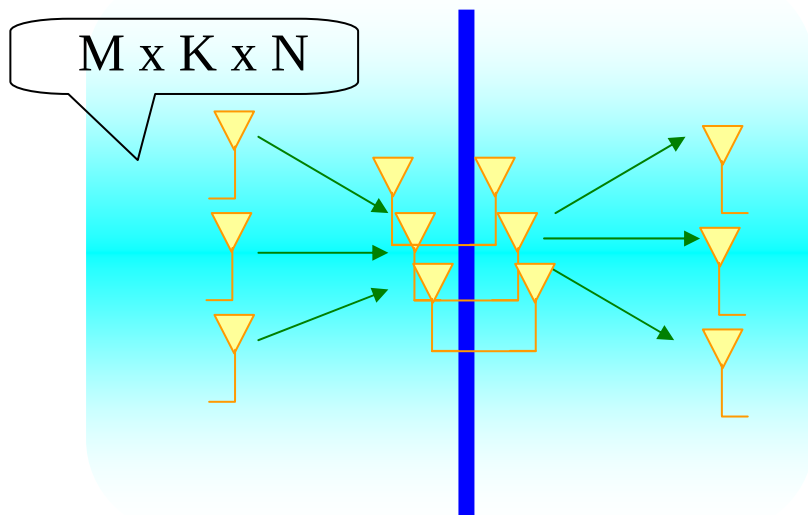
Single-antenna relay



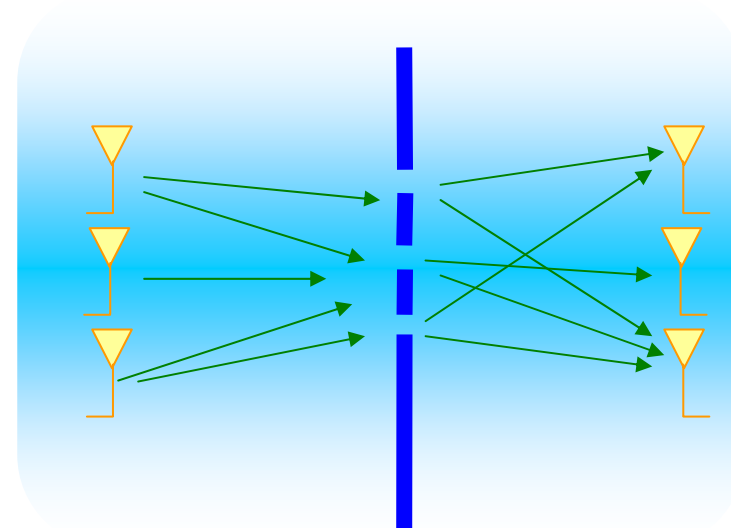
Keyhole environment



Multi-antenna relay

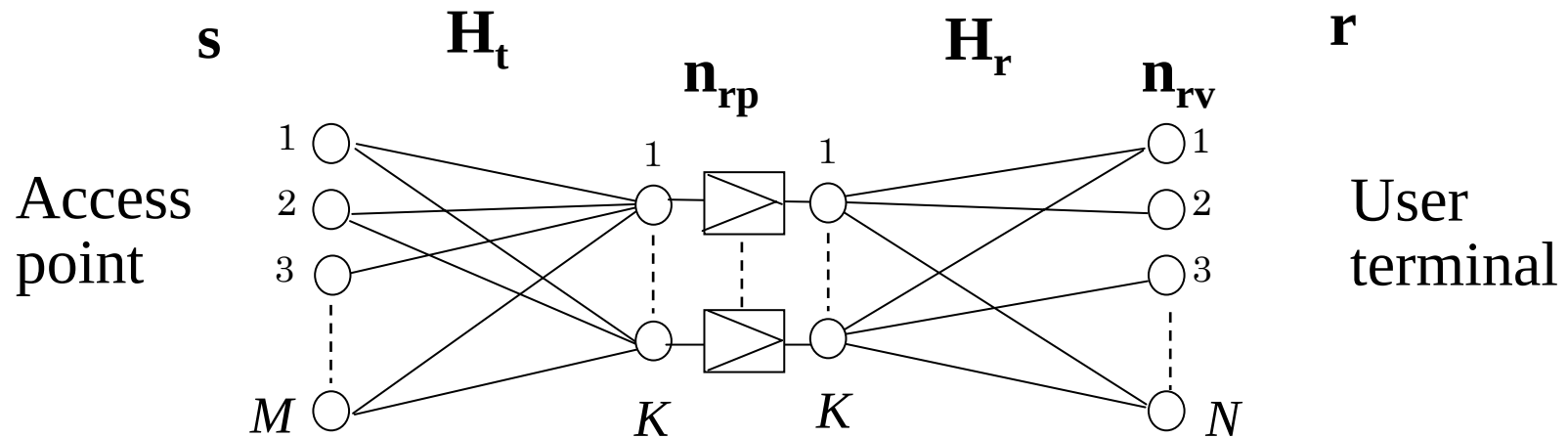


Multi-keyhole environment

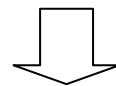




Channel Expression of Radio-Relay Systems



Received signal $\mathbf{r} = \mathbf{H}_r \mathbf{G}(\mathbf{H}_t \mathbf{s} + \mathbf{n}_{rp}) + \mathbf{n}_{rv}$



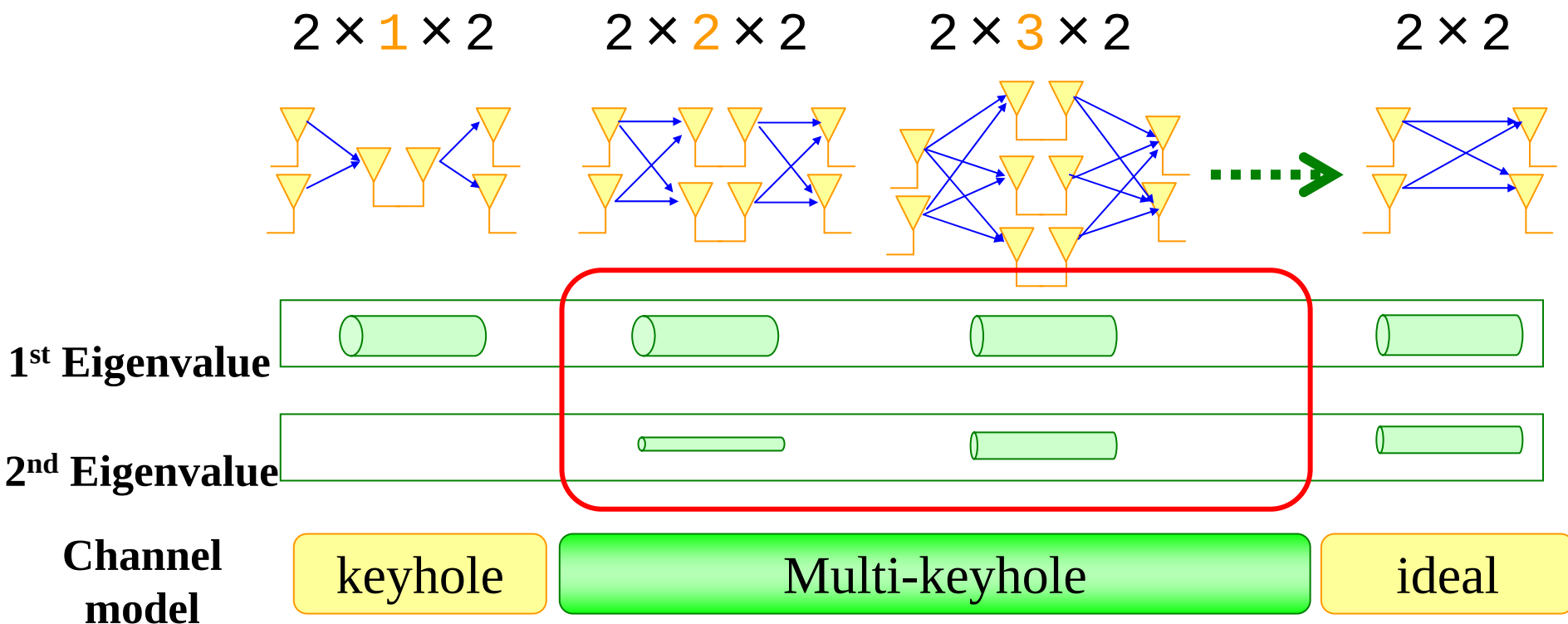
Equivalent channel (CSI)
when the effect
of thermal noise power
in RS is negligible.

$$\mathbf{H}_e = \frac{1}{\sqrt{K}} \mathbf{H}_r \mathbf{H}_t$$

Hereafter, we discuss
PDFs of eigenvalues
for $\mathbf{R} \equiv \mathbf{H}_e \mathbf{H}_e^H$

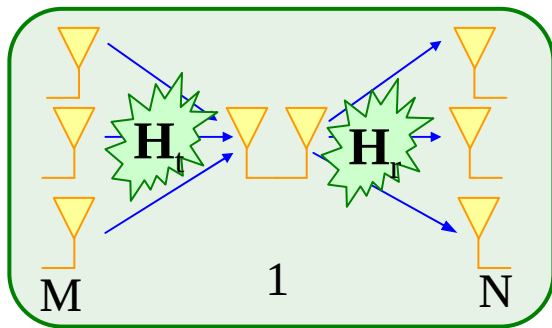


Eigenvalues in MIMO repeater system



Rayleigh fading environment with i.i.d.

PDF of the largest eigenvalue in single keyhole environment



Channel response matrix

$$\mathbf{H}_e = \mathbf{H}_r \mathbf{H}_t$$

$$= \begin{bmatrix} h_{11}^r & h_{21}^r & \cdots & h_{N1}^r \end{bmatrix}^T \begin{bmatrix} h_{11}^t & h_{12}^t & \cdots & h_{1M}^t \end{bmatrix}$$

$$\lambda_1 = \text{Trace}(\mathbf{H}_e \mathbf{H}_e^H)$$

The p.d.f. of eigenvalue

$$p(\lambda_1) = \int_0^\infty \frac{1}{u} p_{2N}(u) p_{2M}\left(\frac{\lambda_1}{u}\right) du$$

$$= \frac{2\lambda_1^{\frac{N+M}{2}-1}}{\Gamma(N)\Gamma(M)} K_{M-N}(2\sqrt{\lambda_1})$$

(K_v : v-th order modified Bessel function of the second kind)

p_{2N}

p_{2M}

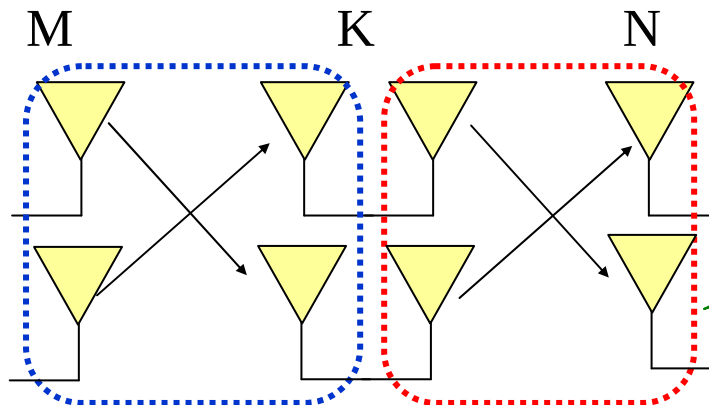
central chi-square
distributed with $2N$
degrees of freedom

central chi-square
distributed with $2M$
degrees of freedom



Approximated PDF of the largest eigenvalue

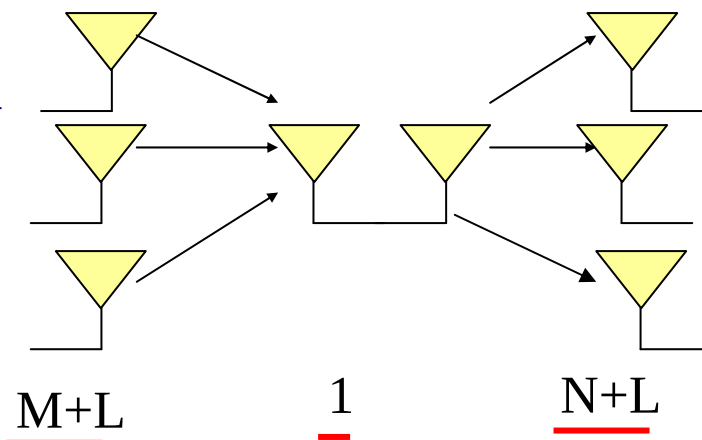
Before the
transformation



linked to the theory of space
diversity at transmission and
reception sides, respectively

Converting the number of
repeater antennas to 1

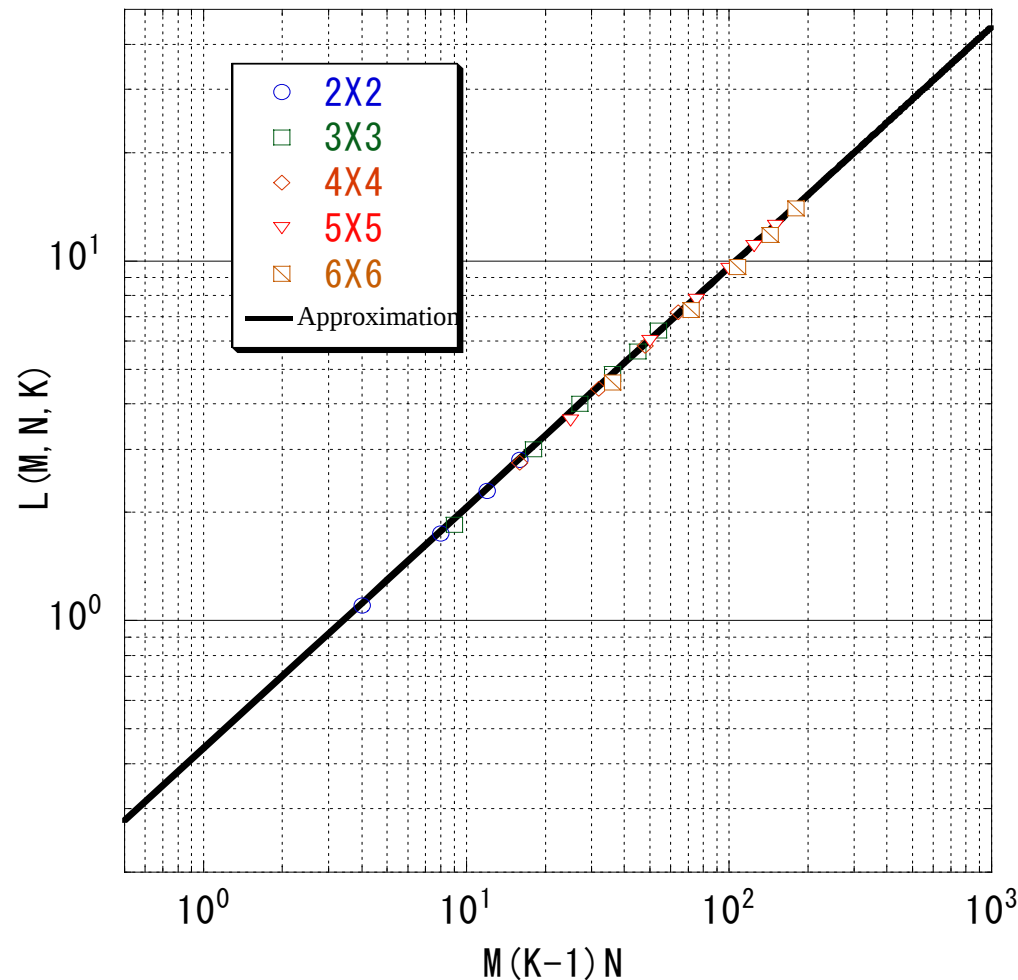
After the
transformation



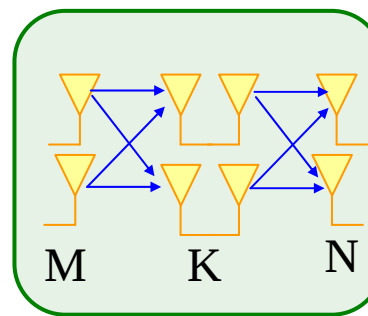
It results to p.d.f. of keyhole

introduction of the number
of effective increment
antennas L

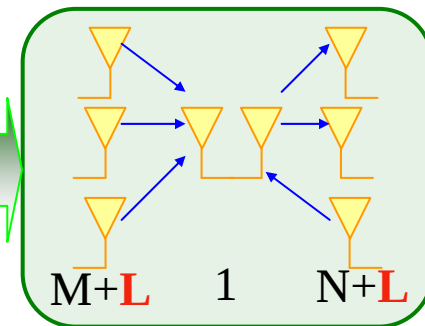
Introduction of the number of effective increment antennas L



Before the transformation



After the transformation



the number of effective increment antennas L

$$L(M, N, K) = \alpha(NM(K-1))^\beta$$

$$\alpha = 0.4343, \beta = 0.6681$$

■ *It is the reasonable approximation.*

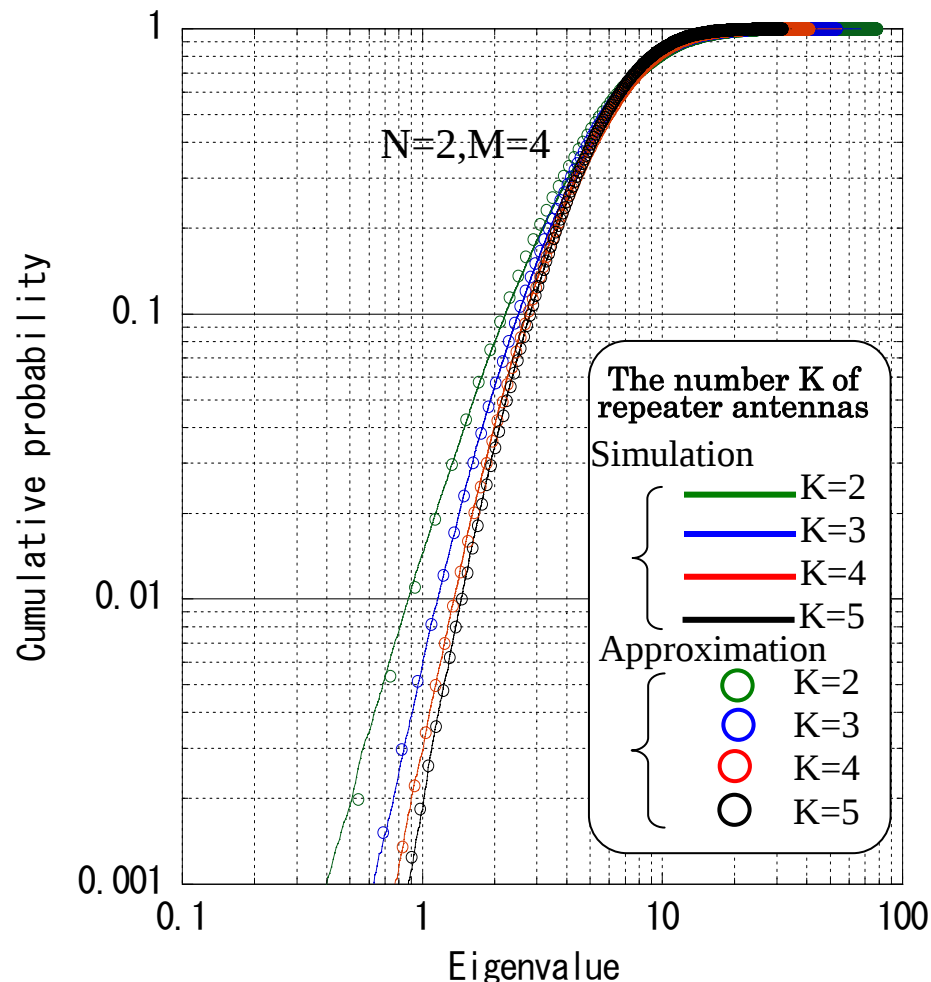
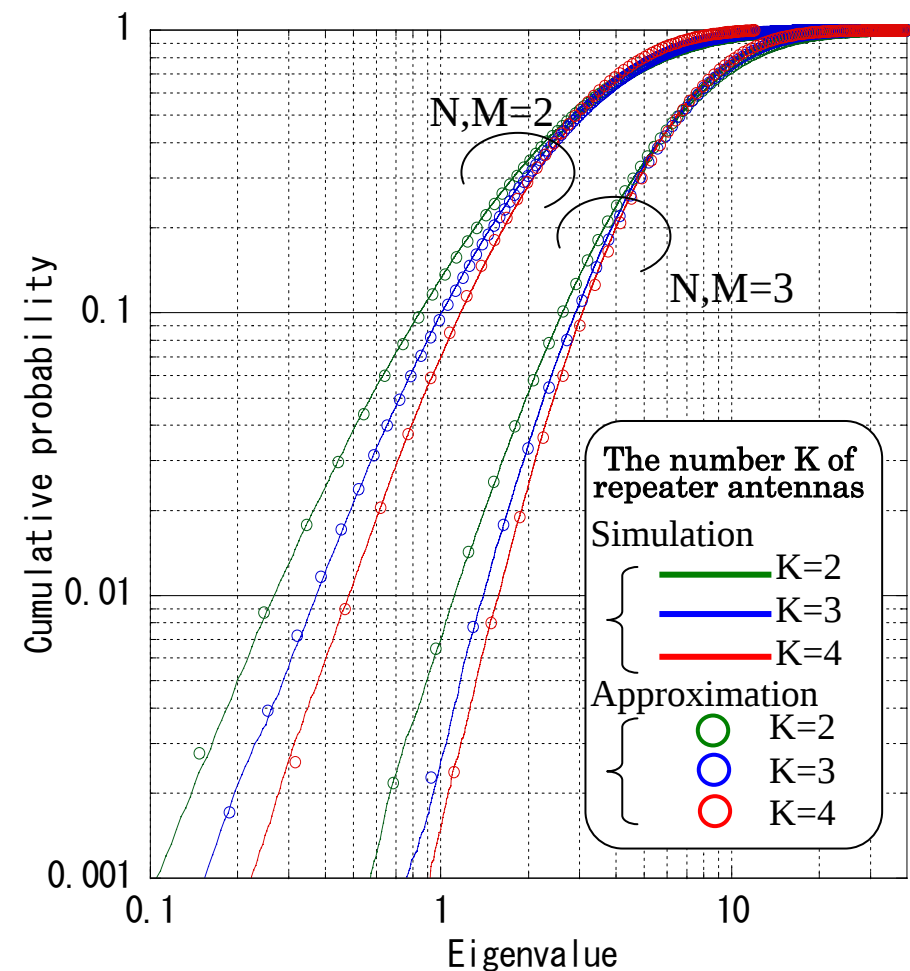


Approximation equation of p.d.f. of largest eigenvalue for MxKxN case

$$p(\lambda_1) = \frac{2 \left(\frac{\lambda_1}{\Lambda} \right)^{\frac{N+M+2L}{2}-1}}{\Lambda \Gamma(N+L) \Gamma(M+L)} K_{M-N} \left(2 \sqrt{\frac{\lambda_1}{\Lambda}} \right)$$

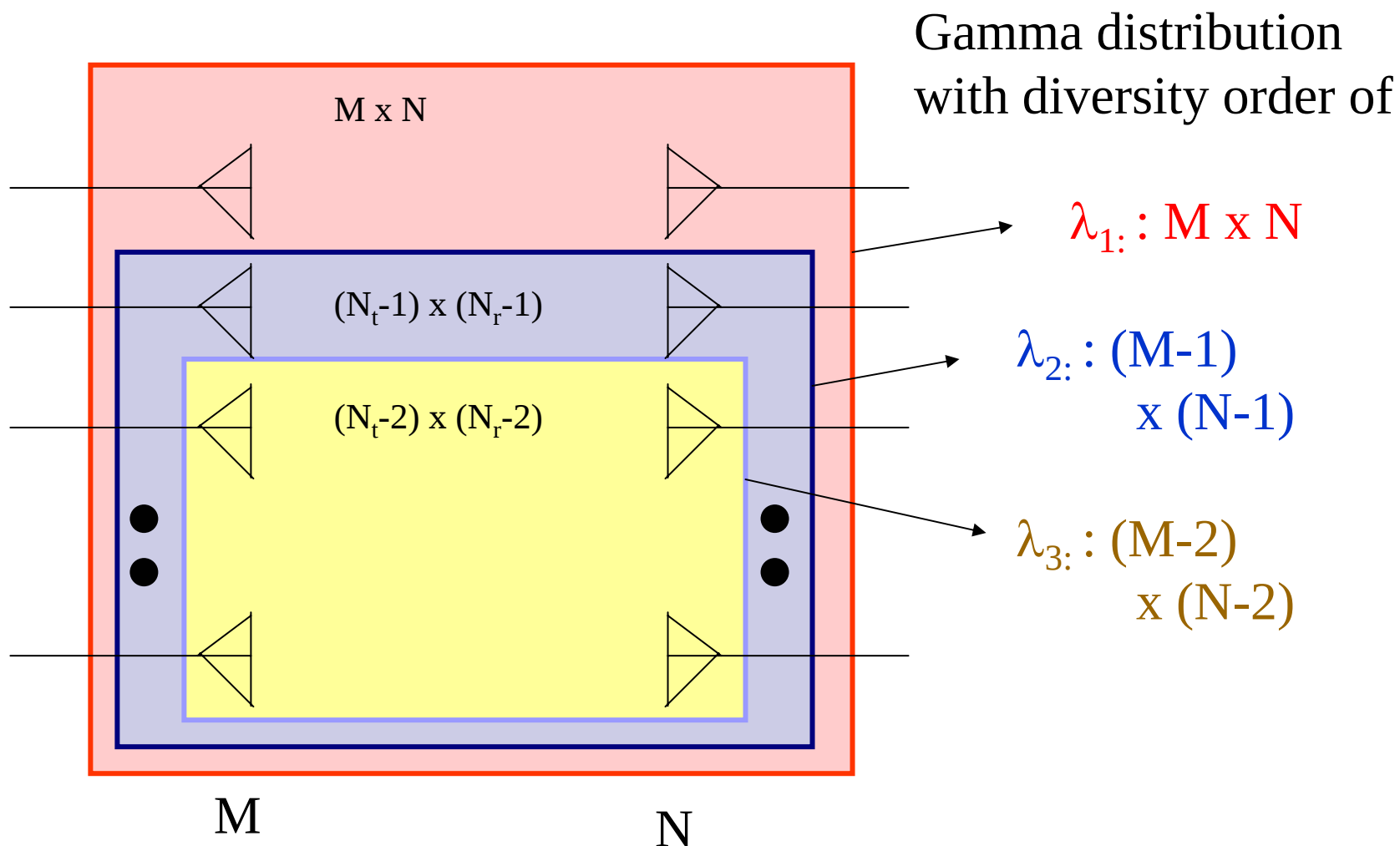
Λ : Power adjustment factor (average value of eigenvalue)

Comparison between simulated values and calculated values ~the largest eigenvalue~

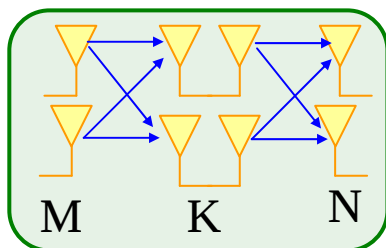




PDF of each Eigenvalue in i.i.d. Environment (approximated)

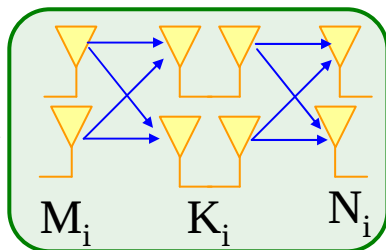


Approximate Formulas of PDFs for All Eigenvalues



Equivalent transformation:
Reduce each antenna number by i-1

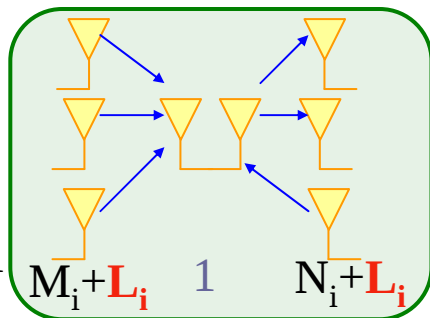
$$\begin{aligned} M_i &= M - i + 1 \\ K_i &= K - i + 1 \\ N_i &= N - i + 1 \end{aligned}$$



$$p(\lambda_i) = \frac{2 \left(\frac{\lambda_i}{\Lambda_i} \right)^{\frac{N_i + M_i + 2L_i}{2} - 1}}{\Lambda_i \Gamma(N_i + L_i) \Gamma(M_i + L_i)} K_{M_i - N_i} \left(2 \sqrt{\frac{\lambda_i}{\Lambda_i}} \right)$$

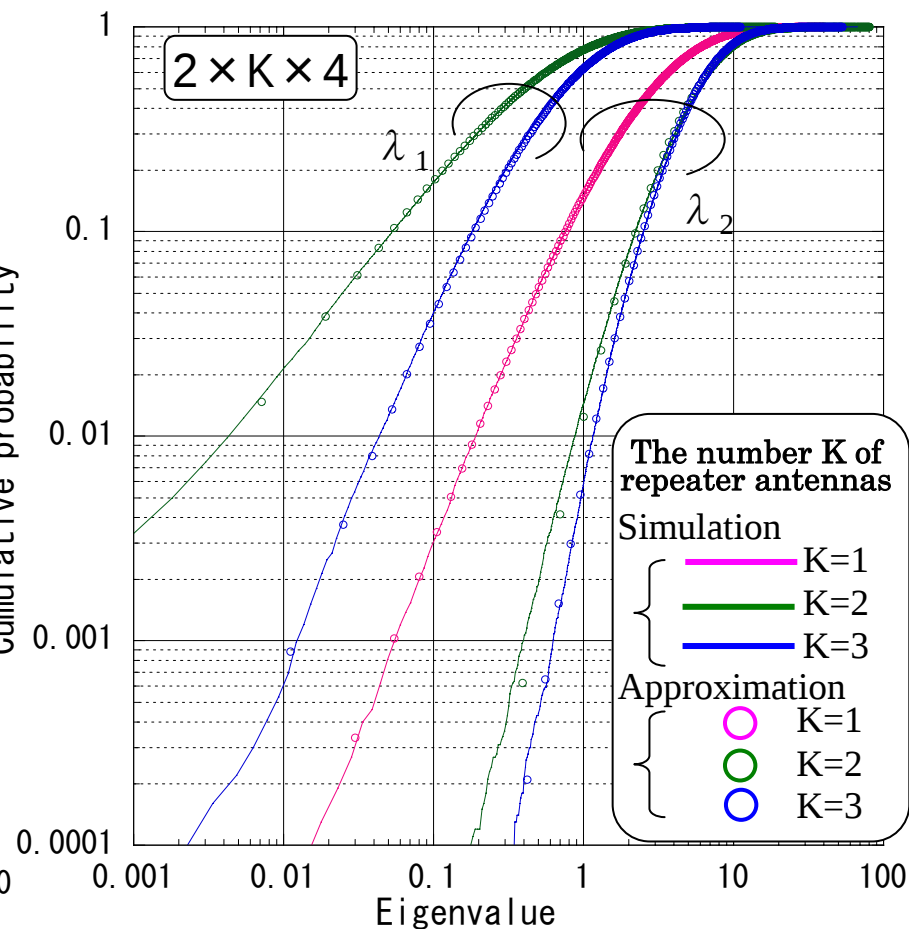
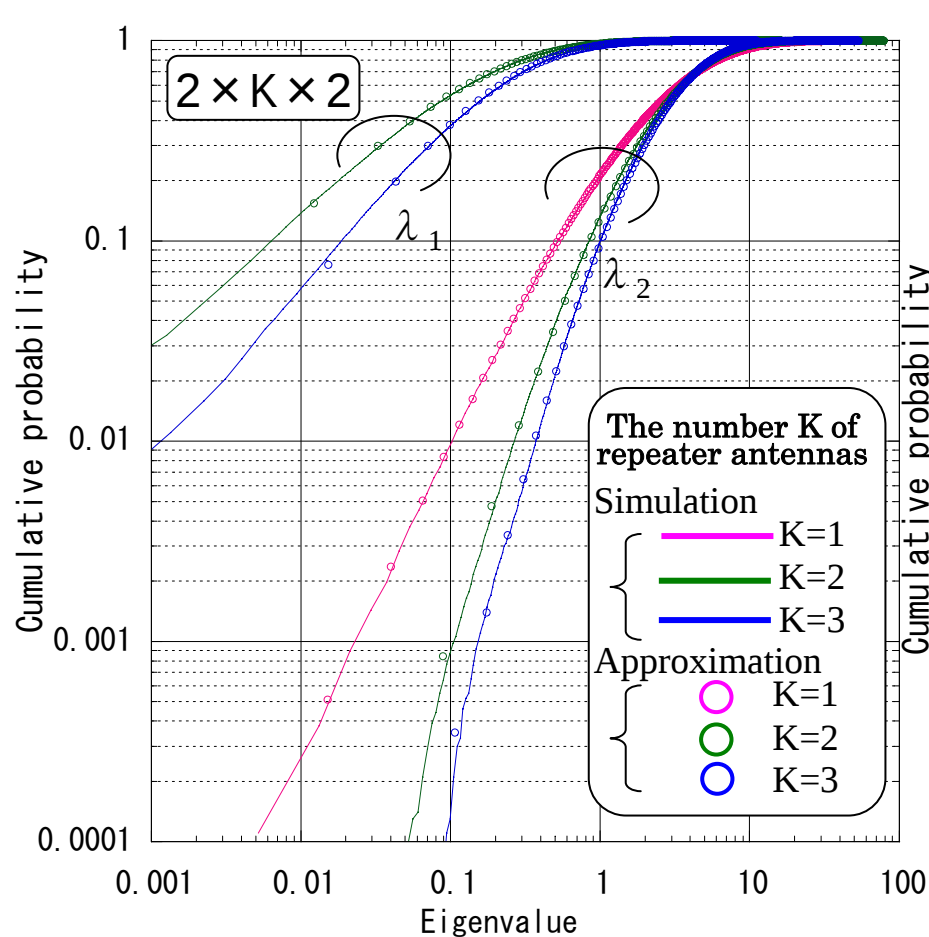
Similar way when obtaining the largest eigenvalue for (M-i+1) x (K-i+1) x (N-i+1)

$$L_i = L - i + 1$$





Comparison between simulated values and calculated values ~all eigenvalues~





Conclusion

- MIMO radio-relay system (or MIMO repeater system) which can expand service area to isolated areas is introduced.
- For the system designing, channel model is important. A channel model named multi-keyhole model for this purpose is presented.
- An empirical calculation method for PDFs of eigenvalues in the channel is developed.
- The proposed scheme realizes very accurate estimate of the PDFs.
- We are ready to evaluate digital transmission characteristics of radio-relay systems using the proposed channel model.